PID Control with Intelligent Compensation

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Neural Control -Compnesation

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PID control (Proportional-Integral-Derivative control) The error between a desired set-point and a measured process variable

$$e = y^* - y$$

The control signal to the process to minimize the error is

$$u = K_{p}e + K_{i}\int_{0}^{t}e\left(au
ight)d au + K_{d}\dot{e}$$

Introduction

- Proportional (P): If the error is large, the controller output will also be large.
- Integral (I): It helps to eliminate steady-state errors.
- Derivative (D): It helps to anticipate future errors and reduce overshoot.



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Neural compensator



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Motivations

PID (model-free) + Intelligent control (model-free) Performance + Stability

• Improved Performance

- Enhanced Robustness:
- Increased Flexibility



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Stability of PD

The dynamics of a robot

$$M(q) \ddot{q} + C(q, \dot{q}) \dot{q} + g(q) + f(\dot{q}) = \tau$$

The classical industrial PD law is

$$au = \mathcal{K}_p(q^d-q) + \mathcal{K}_d(\dot{q}^d-\dot{q})$$

Theorem

The tracking error

$$e = q^* - q$$

is bounded

$$\|e\|^2 < \infty$$

when

 $K_p > 0, \qquad K_d > 0$

The dynamics of a serial n-link rigid robot manipulator

$$M(q)\ddot{q}+C(q,\dot{q})\dot{q}+G(q)+F(\dot{q})= au$$

Property 1. The inertia matrix is symmetric and positive definite , i.e.

$$m_1 \|x\|^2 \le x^T M(x_1) x \le m_2 \|x\|^2$$
; $\forall x \in R^n$

Property 2. The centripetal and Coriolis matrix is skew-symmetric,

$$x^{T}\left[\dot{M}(q)-2C(q,\dot{q})\right]x=0$$

The tracking error is defined as

$$e=q^d-q$$
, $\dot{e}=\dot{q}^d-\dot{q}$

Normal PD control is

$$au = K_p e + K_d \dot{e}$$

where K_p and K_d are positive definite, symmetric and constant matrices

In regulation case

$$\dot{q}^d=0, \quad \dot{e}=\dot{q}^d-\dot{q}=-\dot{q}, \quad \ddot{e}=-\ddot{q}$$

When G and F are known,

$$\tau = K_p e + K_d \dot{e} + G + F$$

The proposed Lyapunov function is

$$V = \frac{1}{2} \dot{e}^T M \dot{e} + \frac{1}{2} e^T K_{\rho} e$$

The derivative of V is

$$\dot{V} = \dot{e}^T M \ddot{e} + rac{1}{2} \dot{e}^T \dot{M} \dot{e} + \dot{e}^T K_{
ho} e$$

Proof of PD cotnrol: Regulation case

Since
$$M\ddot{e} = -M\ddot{q}$$
 and $C\dot{q} + G + F - \tau = -M\ddot{q}$,
 $M\ddot{e} = C\dot{q} + G + F - \tau$

and

$$\dot{e}^{T}M\ddot{e} = \dot{e}^{T}(C\dot{q} + G + F - \tau)$$
$$= -\dot{e}^{T}C\dot{e} + \dot{e}^{T}(G + F - \tau)$$

So

$$\begin{split} \dot{V} &= -\dot{e}^{T}C\dot{e} + \dot{e}^{T}\left(G + F - \tau\right) + \frac{1}{2}\dot{e}^{T}\dot{M}\dot{e} + \dot{e}^{T}K_{p}e \\ &= \frac{1}{2}\dot{e}^{T}\left[\dot{M} - 2C\right]\dot{e} + \dot{e}^{T}\left(G + F - \tau\right) + \dot{e}^{T}K_{p}e \end{split}$$

From Property 2, $x^{T} \left[\dot{M} - 2C \right] x = 0$ and $\tau = K_{p}e + K_{d}\dot{e} + G + F$,

$$\dot{V} = -\dot{e}^{T} \left(K_{p} e + K_{d} \dot{e} \right) + \dot{e}^{T} K_{p} e = -\dot{e}^{T} K_{d} \dot{e} \leq 0$$

LaSalle

Proof of PD cotnrol: Tracking case

$$\dot{q}^d
eq 0$$
, so $\dot{e}
eq -\dot{q}$, $\ddot{e}
eq -\ddot{q}$, the tracking error as:
 $e=q^d-q, \quad \dot{e}=\dot{q}^d-\dot{q}$
Define

$$r = \dot{e} + \Lambda e$$
, $\Lambda = \Lambda^T > 0$

Lyapunov

$$V = \frac{1}{2}r^T M r$$

The derivative of V is

$$\dot{V} = r^T M \dot{r} + \frac{1}{2} r^T \dot{M} r$$

Using $C\dot{q} + G + F - \tau = -M\ddot{q}$, $-Cr + C\dot{q}^d + C\Lambda e = C\dot{q}$

$$\begin{split} M\dot{r} &= M\ddot{e} + M\Lambda\dot{e} = M\ddot{q}^{d} - M\ddot{q} + M\Lambda\dot{e} \\ &= M\ddot{q}^{d} + C\dot{q} + G + F - \tau + M\Lambda\dot{e} \\ &= -\tau - Cr + \left[M\left(\Lambda\dot{e} + \ddot{q}^{d}\right) + C\left(\Lambda e + \dot{q}^{d}\right) + G + F\right] \end{split}$$

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Proof of PD cotnrol: Tracking case

If the control is

 $\tau = Kr + f + G + F$ where $f = M (\Lambda \dot{e} + \ddot{q}^d) + C (\Lambda e + \dot{q}^d)$, then $M\dot{r} = -Kr - Cr$ $\dot{V} = r^T (-Kr - Cr) + \frac{1}{2}r^T \dot{M}r$ $= -r^T Kr - r^T Cr + \frac{1}{2} \tilde{r}^T \dot{M}r$ From Property 2, $r^T \left[\dot{M} - 2C \right] r = 0$ $\dot{V} = -r^T K r < 0$

But we need kto know

$$f=M\left(\Lambda\dot{e}+\ddot{q}^{d}
ight)+C\left(\Lambda e+\dot{q}^{d}
ight)$$
 , $ightarrow$ M , C , G , F

Theorem

For regulation case $\frac{d}{dt}q^* = 0$, g(q) and $f(\dot{q})$ are known, and $\tau = PD + g(q) + f(\dot{q})$ then $\lim_{t \to \infty} ||e||^2 = 0$ if $K_p > 0$, $K_d > 0$

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Theorem

When only g (q) is known, f
$$\left(\stackrel{\cdot}{q} \right)$$
 is unknown, if the control is $au = \mathsf{PD} + \mathsf{g}\left(q
ight)$

and

$$K_p > 0$$
, $K_d > K_1 > 0$

then

$$\|e\|^2 \leq \bar{d}$$

 \overline{d} is upper bound of the friction $f(\dot{q})$.

PD with neural network compensation

When $g\left(q
ight)$ and $f\left(\dot{q}
ight)$ are known, the control is $au=PD+g\left(q
ight)+f\left(\dot{q}
ight)$

and

$$\lim_{t\to\infty}\|e\|^2=0$$

When g(q) and $f(\dot{q})$ are un known, PD with neural network compensation is

$$u = PD + NN$$

where $NN = W_t \sigma(x)$, and

$$g(q) + f\left(\dot{q}\right) = W_t \sigma(x) + \varepsilon$$

Theorem

If the PD control with neual compensator is

$$au = PD + W_t \sigma(x), \quad K_p > 0, K_d > K_1 > 0$$

where the weight W_t is updated as

$$\frac{d}{dt}W_t = K_w \sigma(x) e^T$$

then the tracking error converges to \bar{d}

 $\|e\|^2 \leq \bar{d}$

here \bar{d} is the upper bound of the nerual approximation error ε , i.e.,

$$g(q) + f(\dot{q}) = W_t \sigma(x) + \varepsilon$$

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If the model is unknown, the PD control is

$$\tau = Kr$$

but

$$\begin{aligned} M\dot{r} &= -\tau - Cr + \left[M \left(\Lambda \dot{e} + \ddot{q}^{d} \right) + C \left(\Lambda e + \dot{q}^{d} \right) + G + F \right] \\ &= -Kr - Cr + f + G + F \end{aligned}$$

Then

$$\dot{V} = -r^T Kr - r^T Cr + \frac{1}{2} \dot{e}^T \dot{M} \dot{e} + r^T (f + G + F)$$
$$= -r^T Kr + r^T (f + G + F)$$

Estimation of
$$r^{T} (f + G + F)$$

 $r^{T} (f + G + F) \leq r^{T} \Lambda r + (f + G + F)^{T} \Lambda^{-1} (f + G + F) \leq r^{T} \Lambda r + \bar{d}$
where \bar{d} is upper bound of $(f + G + F)^{T} \Lambda^{-1} (f + G + F)$
 $\dot{V} \leq -r^{T} (K - \Lambda) r + \bar{d}$

when $K > \Lambda$

$$\|r\|^2_{(K-\Lambda)}\to \bar{d}$$

The PD control with neural compensation

 $\tau = \mathit{Kr} + \mathit{W}\sigma(\mathit{x})$

here

$$W\sigma(x) \approx G + F - f$$

where $x = [q^T, \dot{q}^T, e^T, \dot{e}^T, q^d, \dot{q}^d, \ddot{q}^d]^T$, also
 $G + F - f = W^*\sigma(x) + \eta$

where η is bounded modeling error, $\|\eta\|^2 < \bar{\eta}$

Proof of PD cotnrol with neural compensation

The proposed Lyapunov function is

$$V = \frac{1}{2}r^{T}Mr + \frac{1}{2}tr\left(\tilde{W}^{T}K_{w}^{-1}\tilde{W}\right)$$

where $\tilde{\mathcal{W}}=\mathcal{W}^*-\tilde{\mathcal{W}}-.$ The derivative of it is

$$\dot{V} = r^T M \dot{r} + \frac{1}{2} r^T M r + tr \left[\tilde{W}^T K_w^{-1} \left(\frac{d}{dt} \tilde{W} \right) \right] \\ = -r^T K r + r^T \left(f + G + F \right) + tr \left[\tilde{W}^T K_w^{-1} \left(\frac{d}{dt} \tilde{W} \right) \right]$$

Using neural compensator $\tau = Kr + W\sigma(x)$,

$$\begin{aligned} M\dot{r} &= -Kr - Cr + f + G + F \\ &= -Kr - W\sigma(x) + W^*\sigma(x) + \eta \\ &= -Kr - Cr + \tilde{W}\sigma(x) + \eta \end{aligned}$$

Now

$$\dot{V} = -r^{T}Kr + r^{T}\left(\tilde{W}\sigma\left(x\right) + \eta\right) + tr\left[\tilde{W}^{T}K_{w}^{-1}\left(\frac{d}{dt}\tilde{W}\right)\right]$$

Proof of PD cotnrol with neural compensation

The modeling error is estimated as

$$r^{T}\eta \leq r^{T}\Lambda r + \eta^{T}\Lambda^{-1}\eta \leq r^{T}\Lambda r + \overline{\eta}$$
$$\dot{V} \leq -r^{T}(K-\Lambda)r + \overline{\eta} + tr\left[\tilde{W}^{T}\left(K_{w}^{-1}\left(\frac{d}{dt}\tilde{W}\right) + \sigma(x)r^{T}\right)\right]$$

We let

$$K_{w}^{-1}\left(rac{d}{dt}\tilde{W}
ight)+\sigma(x)r^{T}=0$$

the updating law is

$$\frac{d}{dt}\tilde{W} = \frac{d}{dt}W = -K_w\sigma(x)r^T$$

then

$$\dot{V} \leq -r^{T} (K - \Lambda) r + \overline{\eta}$$

when $K > \Lambda$

$$\|r\|^2_{(K-\Lambda)} o\overline\eta$$

here $\overline\eta\llar d$, because $G+F-f=W^*\sigma(x)+\eta_{{}_{1}}$,

$$g(q) + f\left(\dot{q}\right) = W_t \sigma(V_t x) + \varepsilon$$

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PD with MLP compensation

Theorem

If the PD control with neual compensator is

$$au = PD + W_t \sigma(V_t x), \quad K_p > 0, K_d > K_1 > 0$$

where the weight W_t and V_t are updated as

$$egin{aligned} &rac{d}{dt}W_t = -d_t \mathcal{K}_w \sigma(V_t x) \overline{x}_2^T - d_t \mathcal{K}_w D_\sigma V_v x \overline{x}_2^T \ &rac{d}{dt}V_t = -d_t \mathcal{K}_v \overline{x}_2^T D_\sigma W_t x + d_t l \mathcal{K}_v x^T V_t x \Lambda_3 \ &d_t = \left\{egin{aligned} 0 & \textit{if} & \|\overline{x}_2\|_R^2 \leq rac{\chi^2}{4\lambda_{\min}(\Gamma)} \ 1 & \textit{if} & \|\overline{x}_2\|_R^2 > rac{\chi^2}{4\lambda_{\min}(\Gamma)} \end{aligned}
ight.$$

then the tracking error converges to \bar{d}

$$\left\{ \overline{x}_{2} \mid \left\| e \right\|_{R}^{2} \leq rac{ar{d}}{4 \lambda_{\min}\left(\Gamma
ight)}
ight\}$$

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From PD to PID

$$u = K_{p}e + K_{i}\int_{0}^{t}e\left(\tau\right)d\tau + K_{d}\dot{e}$$



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Image: A matrix

PID control law can be expressed

$$u = K_p e - K_d \dot{e} + z$$

 $\dot{z} = K_i e$

with e(0) = 0. The closed-loop system is

$$\frac{d}{dt} \begin{bmatrix} z \\ e \\ \dot{e} \end{bmatrix} = \begin{bmatrix} K_i e \\ -\dot{e} \\ \ddot{q}^d + M^{-1} \left(C\dot{q} + g - K_p \tilde{q} + K_d \dot{q} - z \right) \end{bmatrix}$$

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Theorem

Consider robot dynamic controlled by linear PID controller, the closed loop system is semi-globally asymptotically stable at the equilibrium,

$$\|e\|^2 \leq rac{\lambda_M(M)}{lpha K}, \qquad K > 0$$

provided that control gains satisfy

$$\lambda_{m} (K_{p}) \geq \frac{3}{2} k_{g} \lambda_{M} (K_{i}) \leq \beta \frac{\lambda_{m}(K_{p})}{\lambda_{M}(M)} \lambda_{m} (K_{d}) \geq \beta + \lambda_{M} (M)$$

Consider robot dynamic

$$M(q)\ddot{q}+C(q,\dot{q})\dot{q}+G(q)+F(\dot{q})=u$$

Neural PID control

$$\tau = K_p e + K_d \dot{e} + z + W \sigma(q, \dot{q})$$

$$\dot{z} = K_i e, \quad z(0) = z_0$$

Image: Image:

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Stability of PID+NN

Theorem

Consider robot dynamic controlled by the neural PID control, the closed loop system is semiglobally asymptotically stable at the equilibrium $x = \left[\xi - \phi\left(q^d\right), e, \dot{e}\right]^T = 0$

$$\|e\| \leq \frac{\lambda_{M}(M)}{\kappa_{c}} \lambda_{M}(\kappa_{i}) \lambda_{m}(\kappa_{p})$$
(1)

provided that control gains satisfy

$$\lambda_{m}(K_{p}) \geq \frac{3}{2}k_{\phi}, \lambda_{M}(K_{i}) \leq \beta \frac{\lambda_{m}(K_{p})}{\lambda_{M}(M)}, \lambda_{m}(K_{d}) \geq \beta + \lambda_{M}(M)$$
(2)

where $\beta = \sqrt{\frac{\lambda_m(M)\lambda_m(K_p)}{3}}$, $k_{\phi} > 0$, and the weight of the neural networks is tuned by

$$\frac{d}{dt}W = -K_w\sigma(q,\dot{q})r^T, \qquad r = \dot{e} + \Lambda e^{-K_w\sigma(q,\dot{q})r^T}$$

$$G + F - f = W^* \sigma(x) + \eta(x)$$

the lower bound of $\int \eta(x) dx$ is k_{ϕ} We construct a Lyapunov function as

$$V = V_{1} + V_{2} + V_{3} + V_{4}$$

$$V_{1} = \frac{1}{6}e^{T}K_{p}e + e^{T}\eta + \frac{3}{2}\eta^{T}K_{p}^{-1}\eta$$

$$V_{2} = \frac{1}{6}e^{T}K_{p}e + e^{T}z + \frac{\alpha}{2}z^{T}K_{i}^{-1}z$$

$$V_{3} = \frac{1}{6}e^{T}K_{p}e - \Lambda e^{T}M\dot{e} + \frac{1}{2}\dot{e}^{T}M\dot{e}$$

$$V_{4} = \int_{0}^{t}\eta dx - k_{\phi} + \frac{\alpha}{2}e^{T}K_{d}e + \frac{1}{2}tr\left(\tilde{W}^{T}K_{w}^{-1}\tilde{W}\right)$$
(3)

Proof of of PID+NN

We first prove V is a Lyapunov function, $V \ge 0$.

$$V_4 = \int_0^t \eta \, dx - k_\phi + \frac{\alpha}{2} e^T K_d e + \frac{1}{2} tr\left(\tilde{W}^T K_w^{-1} \tilde{W}\right) \ge 0$$

$$V_1 = \frac{1}{2} \begin{bmatrix} e \\ \eta \end{bmatrix}^T \begin{bmatrix} \frac{1}{3}K_p & I \\ I & 3K_p^{-1} \end{bmatrix} \begin{bmatrix} e \\ \eta \end{bmatrix}$$

Since $K_p \ge 0$, V_1 is a semi positive definite matrix, $V_1 \ge 0$. When $\alpha \ge \frac{3}{\lambda_m(\kappa_i^{-1})\lambda_m(\kappa_p)}$,

$$V_{2} \geq \frac{1}{2} \left(\sqrt{\frac{1}{3} \lambda_{m} \left(K_{p} \right)} \left\| e \right\| - \sqrt{\frac{3}{\lambda_{m} \left(K_{p} \right)}} \left\| z \right\| \right)^{2} \geq 0$$
 (5)

(4)

Proof of of PID+NN

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Because

$$y^{T}Ax \leq \|y\| \|Ax\| \leq \|y\| \|A\| \|x\| \leq |\lambda_{M}(A)| \|y\| \|x\|$$
(6)
when $\alpha \leq \frac{\sqrt{\frac{1}{3}\lambda_{m}(M)\lambda_{m}(K_{p})}}{\lambda_{M}(M)}$,

$$\lambda_{3} \geq \frac{1}{2} \left(\sqrt{\lambda_{m}(M)} \| \dot{e} \| - \sqrt{\frac{1}{3} \lambda_{m}(K_{p})} \| e \| \right)^{2} \geq 0$$
(7)

$$\sqrt{\frac{1}{3}}\lambda_{m}\left(K_{i}^{-1}\right)\lambda_{m}^{\frac{3}{2}}\left(K_{p}\right)\lambda_{m}^{\frac{1}{2}}\left(M\right) \geq \lambda_{M}\left(M\right) \tag{8}$$

there exists

$$\frac{\sqrt{\frac{1}{3}\lambda_{m}\left(M\right)\lambda_{m}\left(K_{p}\right)}}{\lambda_{M}\left(M\right)} \geq \Lambda \geq \frac{3}{\lambda_{m}\left(K_{i}^{-1}\right)\lambda_{m}\left(K_{p}\right)}$$
(9)

This means if K_p is sufficiently large or K_i is sufficiently small, $V(\dot{q}, \tilde{q}, \tilde{\xi})$ is globally positive definite.

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Using $\frac{d}{dt} \int_0^t \eta = \frac{\partial \int_0^t \eta}{\partial x} \frac{\partial x}{\partial t} = \dot{x}^T \eta$, $\frac{d}{dt} \eta = 0$ and $\frac{d}{dt} \left[e^T \eta \right] = \dot{e}^T \eta$, the derivative of V is

$$\dot{V} = \dot{q}^{T} M \ddot{q} + \frac{1}{2} \dot{q}^{T} M \dot{q} + \dot{e}^{T} K_{p} e + \eta^{T} \dot{q}
+ \dot{e}^{T} \eta + tr \left[\tilde{W}^{T} K_{w}^{-1} \left(\frac{d}{dt} \tilde{W} \right) \right]
+ \Lambda \dot{z} K_{i}^{-1} z + \dot{e}^{T} z + e^{T} \dot{z} - \alpha \left(\dot{e}^{T} M \dot{q} + e^{T} M \dot{q} + e^{T} M \ddot{q} \right)
+ \Lambda e^{T} K_{d} \dot{e}$$
(10)

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Using Skew, the first three terms become

$$-\dot{q}^{T}\eta - \dot{q}^{T}K_{d}\dot{q} + \dot{q}^{T}z + \dot{q}^{T}\eta + \dot{q}^{T}\tilde{W}\sigma$$
(11)

And

$$\dot{V} \leq -\left[\lambda_{m}\left(K_{d}\right) - \alpha\lambda_{M}\left(M\right) - \alpha k_{c} \left\|e\right\|\right] \left\|\dot{q}\right\|^{2} - \left[\alpha\lambda_{m}\left(K_{p}\right) - \lambda_{M}\left(K_{i}\right) - \alpha k_{g}\right] \left\|e\right\|^{2}$$
(12)

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Proof of of PID+NN

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$$\|e\| \le \frac{\lambda_M(M)}{\alpha k_c} \tag{13}$$

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and

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then $\dot{V} \leq 0$, $\|e\|$ decreases. Then (14) is established. Using (8) and $\lambda_m \left(K_i^{-1}\right) = \frac{1}{\lambda_M(K_i)}$, (14) is (2).

 \dot{V} is negative semi-definite. Define a ball Σ of radius $\sigma > 0$ centered at the origin of the state space, which satisfies these condition

$$\Sigma = \left\{ \tilde{q} : \|\tilde{q}\| \le \frac{\lambda_M(M)}{\alpha k_c} = \sigma \right\}$$
(15)

 \dot{V} is negative semi-definite on the ball Σ .

There exists a ball Σ of radius $\sigma > 0$ centered at the origin of the state space on which $\dot{V} \leq 0$.

The origin of the closed-loop equation (??) is a stable equilibrium. Since the closed-loop equation is autonomous, we use La Salle's theorem. Define Ω as

$$\Omega = \left\{ x\left(t\right) = \left[\tilde{q}, \dot{q}, \tilde{\xi}\right] \in R^{3n} : \dot{V} = 0 \right\} \\ = \left\{ \tilde{\xi} \in R^{n} : \tilde{q} = 0 \in R^{n}, \dot{q} = 0 \in R^{n} \right\}$$
(16)

From (10), $\dot{V} = 0$ if and only if $\tilde{q} = \dot{q} = 0$. For a solution x(t) to belong to Ω for all $t \ge 0$, it is necessary and sufficient that $\tilde{q} = \dot{q} = 0$ for all $t \ge 0$.

Therefore it must also hold that $\ddot{q} = 0$ for all $t \ge 0$. We conclude that from the closed-loop system (??), if $x(t) \in \Omega$ for all $t \ge 0$, then

$$\begin{aligned} \phi\left(q,\dot{q}\right) &= \phi\left(q^{d},0\right) = \tilde{\xi} + \phi\left(q^{d},0\right) \\ \vdots \\ \tilde{\xi} &= 0 \end{aligned}$$
 (17)

implies that $\tilde{\xi} = 0$ for all $t \ge 0$. So $x(t) = [\tilde{q}, \dot{q}, \tilde{\xi}] = 0 \in \mathbb{R}^{3n}$ is the only initial condition in Ω for which $x(t) \in \Omega$ for all $t \ge 0$.

Finally, we conclude from all this that the origin of the closed-loop system (??) is locally asymptotically stable. Because $\frac{1}{\alpha} \leq \lambda_m \left(K_i^{-1}\right) \lambda_m \left(K_p\right)$, the upper bound for $\|\tilde{q}\|$ can be

$$\|\boldsymbol{e}\| \leq \frac{\lambda_{M}(M)}{k_{c}} \lambda_{M}(K_{i}) \lambda_{m}(K_{p})$$
(18)

It establishes the semiglobal stability of our controller, in the sense that the domain of attraction can be arbitrarily enlarged with a suitable choice of the gains. Namely, increasing K_p the basin of attraction will grow.

Closed-loop tuning of PID control

Since the robot dynamic is not stable in open loop, it is impossible to send step commands to all joints of the robot to tune PID gains., we use

$$PD_{1} = K_{p}e + K_{d}\dot{e}$$
$$PID_{2} = K_{p}e + K_{d}\dot{e} + K_{i}\int_{0}^{t}e\left(\tau\right)d\tau$$

The closed-loop system with PD_1 is stable. If we apply PID_2 to the closed-loop system

$$M(q)\ddot{q} + C(q,\dot{q})\dot{q} + \tilde{g}(q) + f(\dot{q}) - PD_1 = PID_2$$

The total control torque to the robot is

$$\tau = \textit{PID}_2 + \textit{PD}_1$$

if we tune PID controllers *m* times,

$$M\left(q
ight)\ddot{q}+C\left(q,\dot{q}
ight)\dot{q}+g\left(q
ight)+f\left(\dot{q}
ight)=\sum_{j=1}^{m}PID_{j}$$

PID gains are linear independent,

The refinement of PID is the same as adding a new PID controller. For $\ensuremath{\textit{PID}_3}$

Closed-loop tuning od PID control



Figure: PID tuning scheme.

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PID

$$u = K_{p}e + K_{i} \int_{0}^{t} e(\tau) d\tau + K_{d}\dot{e}$$

= $K_{p} (q^{d} - q) + K_{i} \int_{0}^{t} e(\tau) d\tau + K_{d} (\dot{q}^{d} - \dot{q})$

The first-order filter

$$v\left(s\right) = \frac{bs}{s+a}q\left(s\right)$$

where v(s) is an estimation of \dot{q} The transfer function can be realized by

$$\begin{cases} \dot{x} = -A(x + Bq) \\ v = x + Bq \end{cases}$$

The PID control becomes

$$u = K_{p}e + K_{d}v + z + W\sigma(q)$$

$$\dot{z} = K_{i}e$$

$$\dot{x} = -A(x + Bq)$$

$$v = x + Bq$$

$$\|e\| \leq \frac{\lambda_m(M)}{\alpha k_c} \left[\lambda_m(B - \alpha I) - \frac{1}{2}\lambda_m(A)\right] + \frac{1}{\alpha} \|v\|$$

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PD with velocity observer



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PD with neural network compensation

$$u = PD + NN$$

where

$$\hat{f} = g(q) + f\left(\dot{q}\right) = W_t \sigma(V_t x) + \varepsilon$$

/ fuzzy rules

 $\begin{aligned} \mathsf{R}^{i}: \ \mathsf{IF} \ \left(q_{1} \text{ is } A_{1i}^{1}\right) \ \mathsf{and} \ \left(q_{2} \text{ is } A_{2i}^{1}\right) \ \mathsf{and} \ \cdots \ \left(q_{n} \text{ is } A_{ni}^{1}\right) \ \mathsf{THEN} \ \hat{f} \text{ is } B_{1i} \\ \hat{f} &= \hat{W}_{t} \Phi\left(s\right) \\ \phi_{i}^{p} &= \prod_{j=1}^{n} \mu_{A_{ji}^{p}} / \sum_{i=1}^{l} \prod_{j=1}^{n} \mu_{A_{ji}^{p}}, \end{aligned}$

$$\begin{split} \frac{d}{dt}\hat{W}_{p} &= -K_{w}Z_{p}r_{1}^{T} \\ \frac{d}{dt}c_{ji}^{p} &= -2k_{c}z_{i}^{p}\frac{\hat{w}_{pi}-\hat{y}_{p}}{b_{p}}\frac{s_{j}-c_{ji}^{p}}{\left[\sigma_{ji}^{p}\right]^{2}}r_{1,p}^{T} \\ \frac{d}{dt}\sigma_{ji}^{p} &= -2k_{b}z_{i}^{p}\frac{\hat{w}_{pi}-\hat{y}_{p}}{b_{p}}\frac{\left(s_{j}-c_{ji}^{p}\right)^{2}}{\left[\sigma_{ji}^{p}\right]^{3}}r_{1,p}^{T} \end{split}$$

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PD control with fuzzy compensation and high-gain observer



Figure: PD control with fuzzy compensation

PD+SM

$$\tau = K_p e + K_d \dot{e} + K_s sgn(e)$$

The uncertainty is f + G + F. We have to select a very big sliding mode gain, such that

 $K_s > \bar{d}$

$$\begin{split} u &= K_{p}e + K_{d}\dot{e} + W\sigma(x) + K_{s}sgn(e)\\ G &+ F + f = W^{*}\sigma(x) + \eta\\ K_{s} &> \bar{\eta} << \bar{d} \end{split}$$

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PD control with sliding mode control and NN

Theorem

If the updating law for the weights of neural networks is

$$\frac{d}{dt}W = K_w \sigma(x) r_t^T \tag{19}$$

the sliding mode gain

 $K_s > \bar{\eta}$

 $ar\eta$ is the upper bound of the neural approximation error

$$f + G + F = W\sigma(x) + \bar{\eta}$$
, $\|\eta\|^2 \le \bar{\eta}$

then the neural sliding mode PD control forces the tracking error asymptotic stability,

$$\lim_{t \to \infty} r_t = 0 \tag{20}$$

PD control with sliding mode control and NN

SERIAL Neural sliding mode PID control (PID+NN+SMC)



Figure: Tracking error with NN and SMC

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PID control with sliding mode control and NN

$$u = K_p e + K_i \int_0^t e(\tau) d\tau + K_d \dot{e} + W\sigma(x) + K_s sgn(r)$$
$$u = u_1 + (1 - s_t) u_2$$
$$s_t = \begin{cases} 1 & \text{if } ||r||_Q^2 \ge \bar{\eta} \\ 0 & \text{if } ||r||_Q^2 < \bar{\eta} \end{cases}$$

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Theorem

If the sliding mode satisfies

$$K_s > \bar{\eta}$$

where $\bar{\eta}$ is the upper bound of the neural modeling error, then the neural sliding mode PD/PID control with the neural training law

$$\frac{d}{dt}W = s_t K_w \sigma(x) r^T$$

make the tracking error r is stable, and it converge to zero in finite time,

$$||r|| \rightarrow 0$$

Stability of PD and PID

- $\bigcirc \rightarrow \text{Tuning of PID}$
- $\bigcirc \rightarrow \text{PID}+\text{NN}$
- $\bigcirc \rightarrow \mathsf{PID} + \mathsf{Fuzzy}$
- PD→Velocity observer
- $PID+NN \rightarrow PD+NN+SM$

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PID Control with Intelligent Compensation for Exoskeleton Robots



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