

Lecture 13

Dynamic Regulator

13.1 Full order linear dynamic controller

The suggested minimization strategy for quasi-Lipschitz nonlinear output control systems is combined with full-order linear dynamic output controllers in this lecture. In general, we select the controller parameters (gain matrices) that minimize the size of an appealing ellipsoid associated with the closed-loop control system.

Consider the quasi-Lipschitz nonlinear output control system (11.3), (11.4):

$$\left. \begin{aligned} \dot{x}(t) &= Ax(t) + Bu(t) + \xi_x(t, x), \\ x(0) &= x_0 \in \mathbb{R}^n, \\ y(t) &= Cx(t) + \xi_y(t, x). \end{aligned} \right\} \quad (13.1)$$

The admissible feedback control strategies for this system are chosen from the class of *full order linear dynamic controllers* of the following structure

$$\left. \begin{aligned} u(t) &= C_r x_r(t) + D_r y(t), \\ \dot{x}_r(t) &= A_r x_r(t) + B_r y(t), \\ x_r(0) &= x_0^r. \end{aligned} \right\} \quad (13.2)$$

where

$$x_r \in \mathbb{R}^n, A_r \in \mathbb{R}^{n \times n}, B_r \in \mathbb{R}^{n \times k}, D_r \in \mathbb{R}^{m \times k}, C_r \in \mathbb{R}^{m \times n}$$

The control design associated with (13.2) is completely determined by a selection of the matrix

$$\Theta := \begin{bmatrix} D_r & C_r \\ B_r & A_r \end{bmatrix} \in \mathbb{R}^{(n+m) \times (n+k)} \quad (13.3)$$

We call Θ the *dynamic controller matrix*. The closed-loop realization of (13.1) under (13.2) is given by

$$\left. \begin{aligned} \dot{z} &= (A_0 + B_0\Theta C_0)z + D_0\xi_x + B_0\Theta E_0\xi_y, \\ z(t) &:= \begin{pmatrix} x(t) \\ x_r(t) \end{pmatrix} \in \mathbb{R}^{2n}, \quad z(0) = \begin{pmatrix} x_0 \\ x_{r,0} \end{pmatrix}, \end{aligned} \right\} \quad (13.4)$$

where

$$\begin{aligned} A_0 &:= \begin{bmatrix} A & 0_{n \times n} \\ 0_{n \times n} & 0_{n \times n} \end{bmatrix}, \quad B_0 := \begin{bmatrix} B & 0_{n \times n} \\ 0_{n \times m} & I_{n \times n} \end{bmatrix} \\ C_0 &:= \begin{bmatrix} C & 0_{k \times k} \\ 0_{n \times n} & I_{n \times n} \end{bmatrix}, \quad D_0 := \begin{bmatrix} I_{n \times n} \\ 0_{n \times n} \end{bmatrix}, \quad E_0 := \begin{bmatrix} I_{n \times k} \\ 0_{k \times k} \end{bmatrix}. \end{aligned}$$

As in the previous Chapter 12 define the matrix \tilde{G}_B which satisfies the relations

$$\tilde{G}_B \tilde{G}_B^T = \tilde{G}_B^T \tilde{G}_B = I_{2n \times 2n}, \quad \tilde{G}_B B_0 = \begin{bmatrix} 0_{(2n-\tilde{m}) \times \tilde{m}} \\ \tilde{B}_0 \end{bmatrix} \in \mathbb{R}^{2n \times \tilde{m}},$$

$$\tilde{B}_0 \in \mathbb{R}^{\tilde{m} \times \tilde{m}}, \quad \det \tilde{B}_0 \neq 0$$

$$\tilde{G}_B = \begin{pmatrix} B_0^\perp \\ B_0' \end{pmatrix}, \quad \text{where } B_0^\perp = (\text{null}(B_0^T))^\top \text{ and } B_0' = (\text{null}(B_0^\perp))^\top.$$

Following then the same scheme of the analysis as before, we obtain

$$\frac{d}{dt} (\tilde{G}_B z) = (\tilde{G}_B A_0 + \tilde{G}_B B_0 \Theta C_0)z + \tilde{G}_B D_0 \xi_x + \tilde{G}_B B_0 \Theta E_0 \xi_y,$$

or, in new coordinates

$$\tilde{z} := \tilde{G}_B z = \begin{pmatrix} \tilde{z}_1 \in \mathbb{R}^{2n-\tilde{m}} \\ \tilde{z}_2 \in \mathbb{R}^{\tilde{m}} \end{pmatrix}$$

we get

$$\boxed{\begin{aligned} \frac{d}{dt}\tilde{z} &= (\tilde{A}_0 + \begin{bmatrix} 0_{(2n-\tilde{m})\times(n+k)} \\ Y \end{bmatrix}) C_0 \tilde{G}_B^\top \tilde{z} \\ &+ \tilde{G}_B D_0 \xi_x + \begin{bmatrix} 0_{(2n-\tilde{m})\times(n+k)} \\ Y \end{bmatrix} E_0 \xi_y, \end{aligned}} \quad (13.5)$$

where

$$\tilde{A}_0 := \tilde{G}_B A_0 \tilde{G}_B^\top$$

and

$$\boxed{Y := \tilde{B}_0 \Theta \in \mathbb{R}^{\tilde{m}\times(n+k)}}. \quad (13.6)$$

13.2 Main result on the attractive ellipsoid for a dynamic controller

Now we are ready to formulate the principle result concerning the attractive ellipsoid guaranteed by the dynamic controller (13.2).

Theorem 13.1 *If the matrices*

$$P_1 \in \mathbb{R}^{(2n-\tilde{m})\times(n-\tilde{m})}, \quad P_2 \in \mathbb{R}^{\tilde{m}\times\tilde{m}}, \quad Y \in \mathbb{R}^{\tilde{m}\times(n+k)}$$

and positive numbers $\alpha, \varepsilon > 0$ satisfy the following matrix inequalities

$$0 > W_{\alpha,\varepsilon} = \begin{bmatrix} P_1 B_0^\perp A (B_0^\perp)^\top + [P_1 B_0^\perp A (B_0^\perp)^\top]^\top + \alpha P_1 + \varepsilon (\tilde{Q}_x + \tilde{Q}_y) & P_1 B_0^\perp A (B_0')^\top + [P_2 B_0' A (B_0^\perp)^\top]^\top + [Y C (B_0^\perp)^\top]^\top & P_1 B_0^\perp & 0_{(2n-\tilde{m})\times k} \\ [P_1 B_0^\perp A (B_0')^\top]^\top + P_2 B_0' A (B_0^\perp)^\top + Y C (B_0^\perp)^\top & P_2 B_0' A (B_0')^\top + [P_2 B_0' A (B_0')^\top]^\top + [Y C (B_0')^\top]^\top + Y C (B_0')^\top + \alpha P_2 & 0_{\tilde{m}\times n} & P_2 B_0' + P_2 Y \\ (P_1 B_0^\perp)^\top & 0_{n\times\tilde{m}} & -\varepsilon I_{n\times n} & 0_{n\times k} \\ 0_{(2n-\tilde{m})\times k} & (P_2 B_0')^\top + (P_2 Y)^\top & 0_{k\times n} & -\varepsilon I_{k\times k} \end{bmatrix}, \quad (13.7)$$

then the ellipsoid

$$\mathcal{E}_0(\tilde{P}_{attr}) := \{\tilde{z} \in \mathbb{R}^{2n} : \tilde{z}^T \tilde{P}_{attr} \tilde{z} < 1\}, \quad \tilde{P} = \begin{bmatrix} P_1 & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 \end{bmatrix}$$

is attractive for the system (13.5) with

$$\tilde{P}_{attr} = \frac{\alpha}{\varepsilon(c_{0,x} + c_{0,y})} \tilde{P}$$

and the control matrix Θ given by

$$\Theta = \tilde{B}_0^{-1} Y.$$

Proof. Consider the quadratic function of the form

$$V(\tilde{z}) = \tilde{z}^T \tilde{P} \tilde{z}, \quad \tilde{P} = \begin{bmatrix} P_1 & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 \end{bmatrix},$$

$$P_1 \in \mathbb{R}^{(2n-\tilde{m}) \times (2n-\tilde{m})}, \quad P_2 \in \mathbb{R}^{\tilde{m} \times \tilde{m}},$$

and calculate its total derivative along the trajectories of the system (13.5):

$$\begin{aligned} \dot{V}(\tilde{z}) &= 2\tilde{z}^T \tilde{P} \frac{d}{dt} \tilde{z} = \tilde{z}^T \left(\tilde{P} \tilde{A}_0 + \tilde{A}_0^T \tilde{P}^T \right) \tilde{z} + \\ &\tilde{z}^T \left(\begin{bmatrix} 0_{(2n-\tilde{m}) \times (n+k)} \\ Y \end{bmatrix} C_0 \tilde{G}_B^T + \tilde{G}_B C_0^T \begin{bmatrix} 0_{(2n-\tilde{m}) \times (n+k)} & Y^T \end{bmatrix} \right) \tilde{z} \quad (13.8) \\ &+ 2\tilde{z}^T \tilde{P} \tilde{G}_B D_0 \xi_x + 2\tilde{z}^T \tilde{P} \begin{bmatrix} 0_{(2n-\tilde{m}) \times (n+k)} \\ Y \end{bmatrix} E_0 \xi_y \end{aligned}$$

Taking into account that

$$\begin{aligned} \tilde{P} \tilde{A}_0 &:= \tilde{P} \tilde{G}_B A_0 \tilde{G}_B^T = \\ &\begin{bmatrix} P_1 & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 \end{bmatrix} \begin{pmatrix} B_0^\perp \\ B'_0 \end{pmatrix} \begin{bmatrix} A & 0_{n \times n} \\ 0_{n \times n} & 0_{n \times n} \end{bmatrix} \begin{pmatrix} B_0^\perp \\ B'_0 \end{pmatrix}^T = \\ &\begin{bmatrix} P_1 & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 \end{bmatrix} \begin{bmatrix} B_0^\perp A & 0_{n \times n} \\ B'_0 A & 0_{n \times n} \end{bmatrix} \left((B_0^\perp)^T \quad (B'_0)^T \right) = \\ &\begin{bmatrix} P_1 & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 \end{bmatrix} \begin{bmatrix} B_0^\perp A (B_0^\perp)^T & B_0^\perp A (B'_0)^T \\ B'_0 A (B_0^\perp)^T & B'_0 A (B'_0)^T \end{bmatrix} = \\ &\begin{bmatrix} P_1 B_0^\perp A (B_0^\perp)^T & P_1 B_0^\perp A (B'_0)^T \\ P_2 B'_0 A (B_0^\perp)^T & P_2 B'_0 A (B'_0)^T \end{bmatrix} \end{aligned}$$

$$\begin{aligned}
 & \begin{bmatrix} [0_{(2n-\tilde{m}) \times (n+k)}] \\ Y \end{bmatrix} C_0 \tilde{G}_B^\top = \\
 & \begin{bmatrix} [0_{(2n-\tilde{m}) \times (n+k)}] \\ Y \end{bmatrix} \begin{bmatrix} C & 0_{k \times k} \\ 0_{n \times n} & I_{n \times n} \end{bmatrix} \begin{pmatrix} (B_0^\perp)^\top & (B'_0)^\top \end{pmatrix} = \\
 & \begin{bmatrix} [0_{(2n-\tilde{m}) \times (n+k)}] \\ Y \end{bmatrix} \begin{bmatrix} C (B_0^\perp)^\top & C (B'_0)^\top \\ 0_{n \times n} & (B'_0)^\top \end{bmatrix} = \\
 & \begin{bmatrix} 0_{n \times n} & 0_{n \times n} \\ Y C (B_0^\perp)^\top & Y C (B'_0)^\top \end{bmatrix}, \\
 \\
 & \tilde{P} \tilde{G}_B D_0 = \begin{bmatrix} P_1 & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 \end{bmatrix} \begin{pmatrix} B_0^\perp \\ B'_0 \end{pmatrix} \begin{bmatrix} I_{n \times n} \\ 0_{n \times n} \end{bmatrix} = \\
 & \begin{bmatrix} P_1 B_0^\perp & 0_{(2n-\tilde{m}) \times \tilde{m}} \\ 0_{\tilde{m} \times (2n-\tilde{m})} & P_2 B'_0 \end{bmatrix}
 \end{aligned}$$

and

$$\tilde{P} \begin{bmatrix} 0_{(2n-\tilde{m}) \times (n+k)} \\ Y \end{bmatrix} E_0 = \begin{bmatrix} 0_{(2n-\tilde{m}) \times \tilde{m}} & 0_{(2n-\tilde{m}) \times k} \\ 0_{\tilde{m} \times k} & P_2 Y \end{bmatrix},$$

we are able to represent (13.8) as

$$\dot{V}(\tilde{z}) = \begin{pmatrix} \tilde{z}_1 \\ \tilde{z}_2 \\ \xi_x \\ \xi_y \end{pmatrix}^\top W_0 \begin{pmatrix} \tilde{z}_1 \\ \tilde{z}_2 \\ \xi_x \\ \xi_y \end{pmatrix}$$

with

$$W_0 = \begin{bmatrix} P_1 B_0^\perp A (B_0^\perp)^\top + [P_1 B_0^\perp A (B_0^\perp)^\top]^\top & P_1 B_0^\perp A (B'_0)^\top + [P_2 B'_0 A (B_0^\perp)^\top]^\top + [Y C (B_0^\perp)^\top]^\top & P_1 B_0^\perp & 0_{(2n-\tilde{m}) \times k} \\ [P_1 B_0^\perp A (B'_0)^\top]^\top + P_2 B'_0 A (B_0^\perp)^\top + Y C (B_0^\perp)^\top & P_2 B'_0 A (B'_0)^\top + [P_2 B'_0 A (B'_0)^\top]^\top + Y C (B'_0)^\top + [Y C (B'_0)^\top]^\top & 0_{\tilde{m} \times n} & P_2 B'_0 + P_2 Y \\ (P_1 B_0^\perp)^\top & 0_{n \times \tilde{m}} & 0_{n \times n} & 0_{n \times k} \\ 0_{k \times (2n-\tilde{m})} & (P_2 B'_0)^\top + (P_2 Y)^\top & 0_{k \times n} & 0_{k \times k} \end{bmatrix}$$

Adding and subtracting the term $-\alpha V(\tilde{z})$ and in the diagonal blocks $-\varepsilon Q_x$ and $-\varepsilon Q_y$ we derive

$$\dot{V}(\tilde{z}) = \begin{pmatrix} \tilde{z}_1 \\ \tilde{z}_2 \\ \xi_x \\ \xi_y \end{pmatrix}^\top W_1 \begin{pmatrix} \tilde{z}_1 \\ \tilde{z}_2 \\ \xi_x \\ \xi_y \end{pmatrix} - \alpha V(\tilde{z}) + \varepsilon (\xi_x^\top \xi_x + \xi_y^\top \xi_y) \quad (13.9)$$

with

$$W_1 = \begin{bmatrix} P_1 B_0^\perp A (B_0^\perp)^\top + [P_1 B_0^\perp A (B_0^\perp)^\top]^\top + \alpha P_1 & P_1 B_0^\perp A (B_0')^\top + [P_2 B_0' A (B_0^\perp)^\top]^\top + [Y C (B_0^\perp)^\top]^\top & P_1 B_0^\perp & 0_{(2n-\tilde{m}) \times k} \\ [P_1 B_0^\perp A (B_0')^\top]^\top + P_2 B_0' A (B_0^\perp)^\top + Y C (B_0^\perp)^\top & P_2 B_0' A (B_0')^\top + [P_2 B_0' A (B_0')^\top]^\top + [Y C (B_0')^\top]^\top + Y C (B_0')^\top + \alpha P_2 & 0_{\tilde{m} \times n} & P_2 B_0' + P_2 Y \\ (P_1 B_0^\perp)^\top & 0_{n \times \tilde{m}} & -\varepsilon I_{n \times n} & 0_{n \times k} \\ 0_{k \times (2n-\tilde{m})} & (P_2 B_0')^\top + (P_2 Y)^\top & 0_{k \times n} & -\varepsilon I_{k \times k} \end{bmatrix}.$$

As in (11.15)

$$\varepsilon (\xi_x^\top \xi_x + \xi_y^\top \xi_y) \leq \varepsilon \left[(c_{0,x} + c_{0,y}) + \tilde{z}_1^\top \left(\tilde{Q}_x + \tilde{Q}_y \right) \tilde{z}_1 \right]$$

$$\tilde{Q}_x = \tilde{G}_B^\top Q_x \tilde{G}_B^\top, \quad \tilde{Q}_y = \tilde{G}_B Q_y \tilde{G}_B^\top$$

In view of this, the ODE (13.9) can be converted into the following differential inequality

$$\dot{V}(\tilde{z}) \leq \begin{pmatrix} \tilde{z}_1 \\ \tilde{z}_2 \\ \xi_x \\ \xi_y \end{pmatrix}^\top W_{\alpha,\varepsilon} \begin{pmatrix} \tilde{z}_1 \\ \tilde{z}_2 \\ \xi_x \\ \xi_y \end{pmatrix} - \alpha V(\tilde{z}) + \varepsilon (c_{0,x} + c_{0,y}) \quad (13.10)$$

where $W_{\alpha,\varepsilon}$ is defined in (13.7). If $W_{\alpha,\varepsilon} < 0$, the result of this theorem follows from the considerations as in the proof of Theorems in the previous lectures. ■

13.3 Optimal parameters of dynamic controller

Notice that matrix $W_{\alpha,\varepsilon}$ contains the term P_2Y , which makes the matrix inequality (13.7) nonlinear one. To deal with LMI constraint recall that the term corresponds to the term $2\tilde{z}_2^\top P_2Y\xi_x$ in the quadratic form representation. Using Λ -inequality (12.16) with $\Lambda = I_{n \times n}$, we get

$$2\tilde{z}_2^\top P_2Y\xi_x \leq \tilde{z}_2^\top (P_2\Lambda P_2^\top) \tilde{z}_2 + \xi_x^\top (Y^\top \Lambda^{-1} Y) \xi_x \leq$$

$$\tilde{z}_2^\top (P_2^\top P_2) \tilde{z}_2 + \xi_x^\top Y^\top Y \xi_x \leq \tilde{z}_2^\top Q_{\tilde{z}_2} \tilde{z}_2 + \xi_x^\top Q_{\xi_x} \xi_x$$

where matrices $Q_{\tilde{z}_2}$ and Q_{ξ_x} satisfy the matrix inequalities

$$P_2^\top P_2 < Q_{\tilde{z}_2}, \quad Y^\top P_2 Y < Q_{\xi_x} Q_{2,y}$$

which by the Schur's complement are equivalent to the following LMI's:

$$\boxed{\begin{bmatrix} Q_{\tilde{z}_2} & P_2 \\ P_2^\top & I_{n \times n} \end{bmatrix} > 0, \quad \begin{bmatrix} Q_{\xi_x} & Y \\ Y^\top & I_{n \times n} \end{bmatrix} > 0} \quad (13.11)$$

Both matrices $Q_{\tilde{z}_2}$ and Q_{ξ_x} naturally can be incorporated into the corresponding block-diagonals of matrix $W_{\alpha,\varepsilon}$ (13.7), obtaining

$$W_{\alpha,\varepsilon} \leq \bar{W}_{\alpha,\varepsilon} \quad (13.12)$$

where

$$\boxed{\begin{array}{cccc} & & & \bar{W}_{\alpha,\varepsilon} = \\ \left[\begin{array}{cccc} P_1 B_0^\perp A (B_0^\perp)^\top + [P_1 B_0^\perp A (B_0^\perp)^\top]^\top + \alpha P_1 + \varepsilon (\tilde{Q}_x + \tilde{Q}_y) & P_1 B_0^\perp A (B_0')^\top + [P_2 B_0' A (B_0^\perp)^\top]^\top + [Y C (B_0^\perp)^\top]^\top & P_1 B_0^\perp & 0_{(2n-\tilde{m}) \times k} \\ [P_1 B_0^\perp A (B_0')^\top]^\top + P_2 B_0' A (B_0^\perp)^\top + Y C (B_0^\perp)^\top & P_2 B_0' A (B_0')^\top + [P_2 B_0' A (B_0')^\top]^\top + [Y C (B_0')^\top]^\top + Y C (B_0')^\top + \alpha P_2 + Q_{\tilde{z}_2} & 0_{\tilde{m} \times n} & P_2 B_0' + P_2 Y \\ (P_1 B_0^\perp)^\top & 0_{n \times \tilde{m}} & -\varepsilon I_{n \times n} + Q_{\xi_x} & 0_{n \times k} \\ 0_{(2n-\tilde{m}) \times k} & (P_2 B_0')^\top + (P_2 Y)^\top & 0_{k \times n} & -\varepsilon I_{k \times k} \end{array} \right] \cdot \end{array}} \quad (13.13)$$

Fulfilling $\bar{W}_{\alpha,\varepsilon} < 0$, in view of (13.12), we guarantee $W_{\alpha,\varepsilon} < 0$.

The optimization procedure for finding the best parameters Θ for the dynamic controller, which provide the minimal size of the attractive ellipsoid, can be formulated now similarly to the problem from lecture 9, namely,

$$\inf_{P_1>0, P_2>0, Q_{z_2}>0, Q_{\xi_x}>0, Y, \alpha>0, \varepsilon>0} [\text{tr}(P_1^{-1}) + \text{tr}(P_2^{-1})],$$

or equivalently,

$$\boxed{\inf_{P_1>0, P_2>0, Q_{z_2}>0, Q_{\xi_x}>0, H_1>0, H_2>0, Y, \alpha>0, \varepsilon>0} [\text{tr}(H_1) + \text{tr}(H_2)],}$$

subject to LMI's constraints (13.13), (13.11) and additionally (13.14)

$$\boxed{\begin{aligned} & \left[\begin{array}{cc} H_1 & I_{(n-\tilde{m}) \times (n-\tilde{m})} \\ I_{(n-\tilde{m}) \times (n-\tilde{m})} & P_1 \end{array} \right] > 0, \\ & \left[\begin{array}{cc} H_2 & I_{\tilde{m} \times \tilde{m}} \\ I_{\tilde{m} \times \tilde{m}} & P_2 \end{array} \right] > 0, \\ & H_1 \in \mathbb{R}^{(n-m) \times (n-m)}, H_2 \in \mathbb{R}^{\tilde{m} \times \tilde{m}}, \end{aligned}} \quad (13.14)$$

obtained from the matrix estimates

$$P_1^{-1} < H_1, P_2^{-1} < H_2$$

by the Schur's complement implementation.

13.4 Exercise

Exercise 13.1 For the system (as in the previous lecture 8)

$$\left. \begin{aligned} \dot{x}_1 &= -x_1 + x_2 + 0.1x_1 \text{sign}(x_2), \\ \dot{x}_2 &= -x_1 + 0.2 \text{sign}(x_1) + u, \\ y &= x_1 + x_2 + \xi_y(t), \\ x_1, x_2 &\in \mathbb{R}, x_0 = (1, 1)^\top, \\ |\xi_y(t)|^2 &\leq c_{0,y} = 0.01. \end{aligned} \right\}$$

design the dynamic controller and find its optimal numerical feedback parameters Θ^* . To compare graphically the obtained results with the previous lecture results.